

GNSS L5/E5 Maximum Likelihood Synchronization Performance Degradation under DME Interferences

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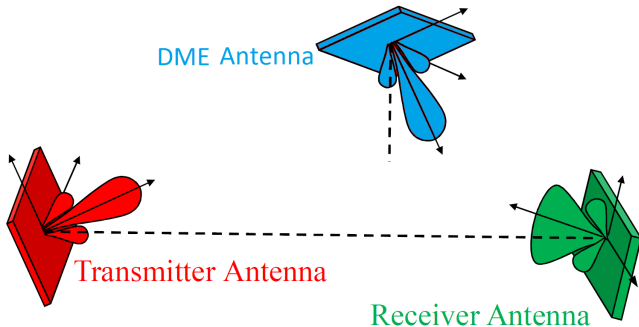


Figure: Illustration of a GNSS system interfered by a DME transmitter.

Previous Results

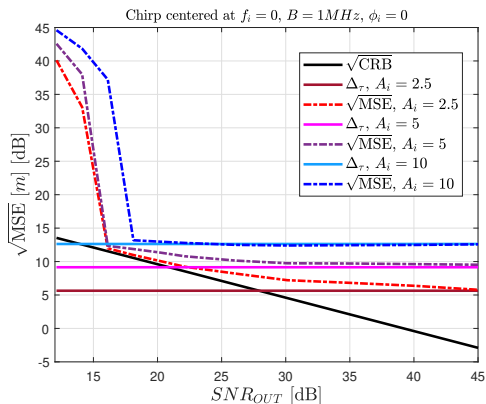


Figure: MSE and bias for the time-delay estimation with a chirp centered at $f_i = 0$, for different jammer amplitudes A_i . Chirp bandwidth 1 MHz and initial jammer phase $\phi = 0$.

- Misspecified CRB

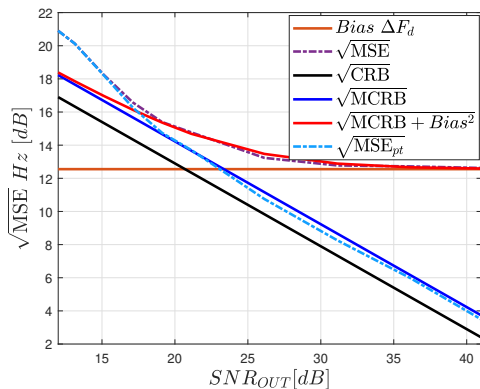


Figure: MSE and bias for the time-delay (left) and Doppler (right) estimation with a chirp centered at $f_i = 0\text{MHz}$, for different jammer amplitudes $A_i = 10$. Chirp bandwidth 2 MHz and initial jammer phase $\phi = 0$.

- Signal Model
 - Correctly Specified Signal Model
 - Misspecified Signal Model
- How to Compute the Bias: Kullback-Leibler Divergence
- How to Compute the MCRB
- MSE Computation and Results for the DME Interference Models

- **Correctly Specified Signal Model [3]:** A GNSS band-limited signal $s(t)$, with bandwidth B , transmitted over a carrier frequency f_c ($\lambda_c = c/f_c$) is considered in this study. The baseband output of the receiver's Hilbert filter can be approximated by,

$$x(t; \boldsymbol{\eta}) = \alpha s(t - \tau) e^{-j2\pi f_c(b(t-\tau))} + l(t) + n(t), \quad (1)$$

with $\boldsymbol{\eta} = (\tau, b)^\top$, $l(t)$ an unknown interference, $n(t)$ a complex white Gaussian noise. The discrete vector signal model is built from $N_1 \leq k \leq N_2$ samples at $T_s = 1/F_s$,

$$\mathbf{x} = \alpha \mathbf{a}(\boldsymbol{\eta}) + \mathbf{n} = \rho e^{j\Phi} \mathbf{a}(\boldsymbol{\eta}) + \mathbf{n} = \alpha \boldsymbol{\mu}(\boldsymbol{\eta}) + \mathbf{l} + \mathbf{n}, \quad (2)$$

$$\mathbf{a}(\boldsymbol{\eta}) = (\dots, s(kT_s - \tau) e^{-j2\pi f_c(b(kT_s - \tau))} + \frac{1}{\alpha} l(kT_s), \dots)^\top. \quad (3)$$

The unknown deterministic parameters can be gathered in vector $\boldsymbol{\epsilon}^\top = (\sigma_n^2, \rho, \Phi, \boldsymbol{\eta}^\top) = (\sigma_n^2, \boldsymbol{\theta}^\top)$, with $\rho \in \mathbb{R}^+$, $0 \leq \Phi \leq 2\pi$. Then, $\mathbf{x} \sim \mathcal{CN}(\alpha \mathbf{a}(\boldsymbol{\eta}), \sigma_n^2 \mathbf{I}_N)$.

- **Misspecified Signal Model [2, 1]:** The misspecified signal model represents the case where the interference is not considered. This nominal case leads to the definition of the misspecified parameter vector $\boldsymbol{\eta}' = [\tau', b']^\top$, and the complete set of unknown parameters $\boldsymbol{\epsilon}'^\top = [\sigma_n^2, \rho', \Phi', \boldsymbol{\eta}'^\top] = [\sigma_n^2, \boldsymbol{\theta}'^\top]$, yielding the following signal model at the output of the Hilbert filter,

$$x'(t; \boldsymbol{\eta}') = \alpha' s(t - \tau') e^{-j2\pi f_c b'(t - \tau')} + n(t) \quad (4)$$

with $\alpha' = \rho' e^{j\Phi'}$ and the discrete vector signal model:

$$\mathbf{x}' = \alpha' \boldsymbol{\mu}(\boldsymbol{\eta}') + \mathbf{n} \quad (5)$$

$$\boldsymbol{\mu}(\boldsymbol{\eta}') = (\dots, s(kT_s - \tau') e^{-j2\pi f_c (b'(kT_s - \tau'))}, \dots)^\top.$$

The misspecified signal model is represented by a pdf denoted as $\mathbf{x}' \sim \mathcal{CN}(\alpha' \boldsymbol{\mu}(\boldsymbol{\eta}'), \sigma_n^2 \mathbf{I}_N)$.

How to Compute the Bias: Kullback-Leibler Divergence

The pdfs of the Correctly Specified and Misspecified Signal Models are [4]:

$$p_{\epsilon}(x; \epsilon) = \frac{1}{\pi^N \sigma_n^{2N}} e^{\frac{-(x - \alpha a(\eta))^H (x - \alpha a(\eta))}{\sigma_n^2}}, \quad (6)$$

$$f_{\epsilon'}(x'; \epsilon') = \frac{1}{\pi^N \sigma_n^{2N}} e^{\frac{-(x - \alpha' \mu(\eta'))^H (x - \alpha' \mu(\eta'))}{\sigma_n^2}}. \quad (7)$$

When we consider a misspecified model and the corresponding MMLE, the estimation of the parameters of interest is biased. Those biased estimated parameters are commonly referred to as pseudo-true parameters. We denote them as $\theta_{pt}^{\top} = [\rho_{pt}, \Phi_{pt}, \tau_{pt}, b_{pt}]$.

The pseudo-true parameters are simply those that give the minimum Kullback-Leibler (KLD) Divergence $D(p_{\epsilon} || f_{\epsilon'})$ between the true and assumed models.

$$\Delta\alpha = \alpha_{pt} - \alpha, \quad \Delta\eta = \eta_{pt} - \eta. \quad (8)$$

How to Compute the MCRB

The MCRB was derived as an extension of the Slepian-Bangs formulas, and it is a combination of two information matrices $A(\boldsymbol{\theta}_{pt})$ and $B(\boldsymbol{\theta}_{pt})$

$$\text{MCRB}(\boldsymbol{\theta}_{pt}) = A(\boldsymbol{\theta}_{pt})^{-1}B(\boldsymbol{\theta}_{pt})A(\boldsymbol{\theta}_{pt})^{-1}, \quad (9)$$

where

$$A(\boldsymbol{\theta}_{pt}) = \frac{2}{\sigma_n^2} \Re \left\{ (\delta \mathbf{m})^H \left(\frac{\partial^2 \alpha_{pt} \boldsymbol{\mu}(\boldsymbol{\eta}_{pt})}{\partial \boldsymbol{\theta}_{pt} \partial \boldsymbol{\theta}_{pt}^\top} \right) \right\} - B(\boldsymbol{\theta}_{pt}),$$
$$B(\boldsymbol{\theta}_{pt}) = \frac{2}{\sigma_n^2} \Re \left\{ \left(\frac{\partial \alpha_{pt} \boldsymbol{\mu}(\boldsymbol{\eta}_{pt})}{\partial \boldsymbol{\theta}_{pt}} \right)^H \left(\frac{\partial \alpha_{pt} \boldsymbol{\mu}(\boldsymbol{\eta}_{pt})}{\partial \boldsymbol{\theta}_{pt}} \right) \right\},$$

$\delta \mathbf{m} \triangleq \alpha \mathbf{a}(\boldsymbol{\eta}) - \alpha_{pt} \boldsymbol{\mu}(\boldsymbol{\eta}_{pt}) = \alpha \boldsymbol{\mu}(\boldsymbol{\eta}) + \mathbf{I} - \alpha_{pt} \boldsymbol{\mu}(\boldsymbol{\eta}_{pt})$ the mean difference between true and misspecified models.

We have derived **closed-expressions** of these formulas that depend only on the signal and interference samples.

MSE Computation and Results

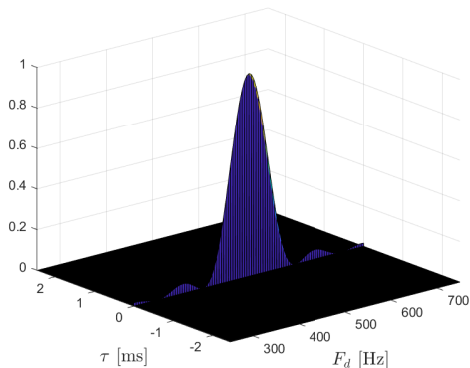


Figure: Ambiguity function of the GPS L5Q signal. The integration time is set to 10 ms, $F_s = 20$ MHz.

MSE Computation and Results

The following scenario is proposed We consider 7 DME signals interfering the GNSS receiver, with $f_i = [-0.5, -0.25, 0.15, 0, 0.15, 0.25, 0.5]$ MHz, and arriving at the receiver $[1.25, 2.5, 3.75, 5, 6.25, 7.5, 8.75]$ ms after the first chip of the GNSS PRN code.

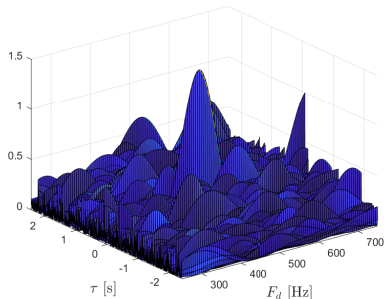
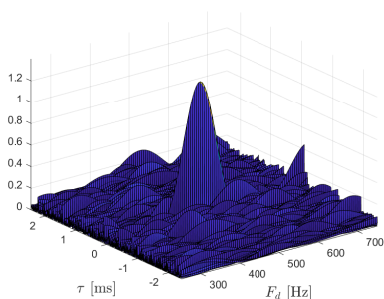


Figure: Ambiguity function of the GPS L5Q signal. The integration time is set to 10 ms, $F_s = 20$ MHz. $I_i = \{30, 32\}$ dB.

MSE Computation and Results

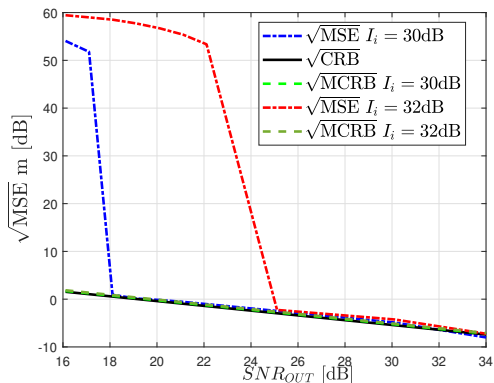


Figure: RMSE of the time-delay of the GPS L5Q signal. The integration time is set to 10 ms, $F_s = 20$ MHz and $I_i = \{30, 32\}$ dB.

MSE Computation and Results

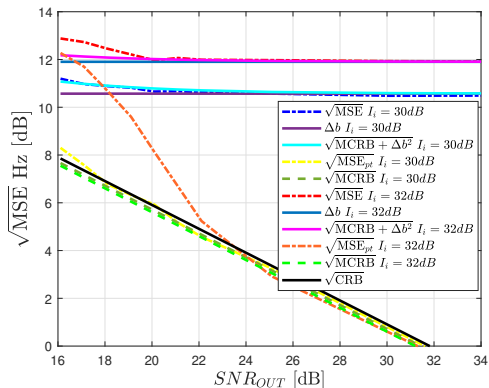


Figure: RMSE of the Doppler of the GPS L5Q signal. The integration time is set to 10 ms, $F_s = 20$ MHz and $I_i = \{30, 32\}$ dB.

MSE Computation and Results

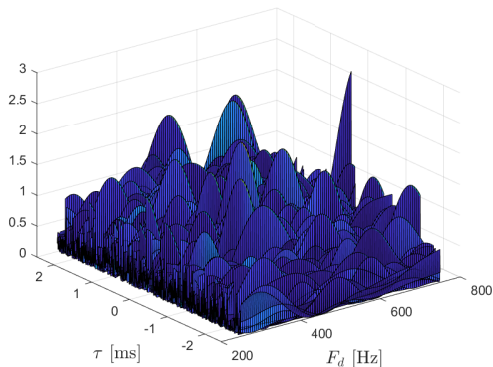


Figure: Ambiguity function of the GPS L5Q signal. The integration time is set to 10 ms, $F_s = 20$ MHz. $I_i = 34$ dB.

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- [2] Hamish McPhee et al. “On the accuracy limits of misspecified delay-Doppler estimation”. In: *Signal Processing* 205 (2023), p. 108872. ISSN: 0165-1684. DOI: <https://doi.org/10.1016/j.sigpro.2022.108872>.
- [3] D. Medina et al. “A New Compact CRB for Delay, Doppler and Phase Estimation - Application to GNSS SPP & RTK Performance Characterization”. In: *IET Radar, Sonar & Navigation* (2020).

- [4] L. Ortega, J. Vilà-Valls, and E. Chaumette. “Theoretical Evaluation of the GNSS Synchronization Performance Degradation under Interferences”. In: *Proceedings of the 35th International Technical Meeting of the Satellite Division of The Institute of Navigation (ION GNSS+ 2022)*. 2022, pp. 3758–3767. DOI: <https://doi.org/10.33012/2022.18564>.

thank you!